

CapLibrate: Self-Calibration of an Energy Harvesting Power Supply with Supercapacitors

Christian Renner and Volker Turau

GI/ITG Workshop on Energy-aware Systems and Methods

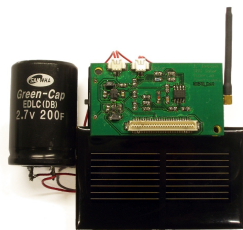
22nd February, 2010



Introduction

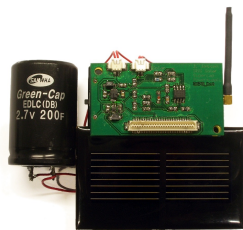
Modern Wireless Sensor Nodes

- Low-Power Components
- Energy Harvesting Power Supply
- Energy Buffer



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Unattended Operation \Rightarrow Power Management

Energy Awareness

Knowledge of available energy \Rightarrow Energy Awareness

Benefits for known task energy costs

- Lifetime prediction
- Efficient use of available energy resources, e.g.,
 - ◆ Energy-aware scheduling
 - ◆ Energy-aware network-wide decisions
 - routing, clustering, ...
- Low risk of energy starvation
- Perpetual operation

SuperCaps – the Good News



- Small and light-weighted
- Long lifetime
- Virtually unlimited charge/discharge cycles
- Simple charging circuit
- Easy estimation of stored energy

$$T_{\text{life}} = \frac{\eta C}{2P_N} (V_C^2 - V_{\text{min}}^2)$$

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The Bad News



Capacity C depends on

- manufacturing variation
- temperature
- age

Circuit efficiency η depends on

- actual DC-DC-efficiency
- manufacturing variation
- part tolerances
- hardware revisions
(different parts)

Manual calibration infeasible in large networks

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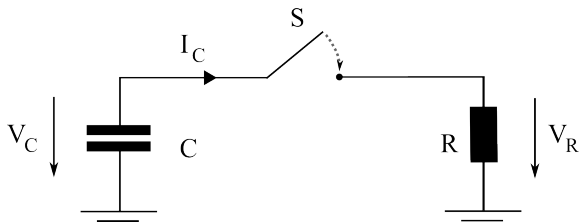
Dish of the Day

Self-Calibration

Theory in Praxis

Analytical Assessment

Resistance Discharging

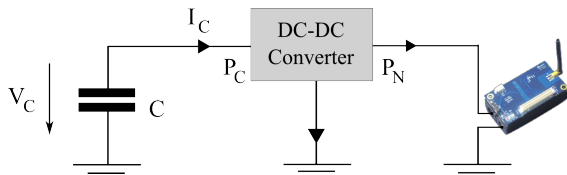


$$V_C = V_{C0} \cdot e^{-\frac{\Delta T}{RC}}$$

Resistance Discharging – Solution

$$C = \frac{\Delta T}{R} \cdot \ln^{-1} \left(\frac{V_{C0}}{V_C} \right)$$

Converter Discharging



$$E_C = E_{\text{out}} \quad \Leftrightarrow \quad \frac{C \cdot (V_{C0}^2 - V_C^2)}{2} = \frac{\Delta T \cdot P_N}{\eta}$$

Converter Discharging – Solution

$$C = \frac{2 \cdot \Delta T \cdot P_N}{\eta \cdot (V_{C0}^2 - V_C^2)}$$



Evaluation

Environment

- Crossbow IRIS node with TinyOS 2.1
- Sampling of V_C via node's ADC
Averaging 4 readings to reduce noise
- Sending result via radio to PC
- Samwha GreenCaps: $C = 25\text{ F}, 100\text{ F}, 200\text{ F}$
- Initial value of $V_C = 2.7\text{ V}$
- Reference voltages $V_{C0} = 1.5\text{ V}, 2.0\text{ V}, 2.5\text{ V}$

Discharging Setups

1. Stand-alone Resistance Discharging

- ◆ $R = 36.1 \Omega$
- ◆ Node powered by battery

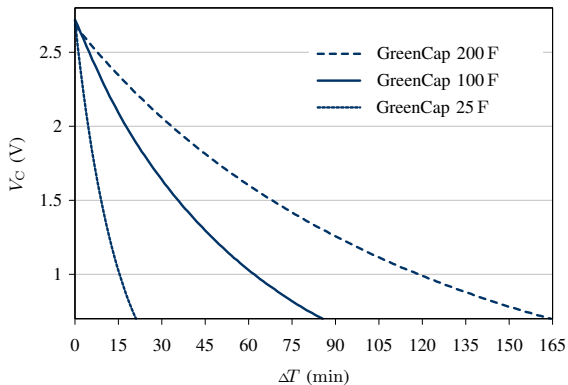
2. Combined Resistance Discharging ▷ Math

- ◆ $R = 35.6 \Omega$, switched via MOSFET
- ◆ Node powered by SuperCap
- ◆ $P_N < 100 \mu\text{W}$ (sleep mode)

3. Converter Discharging

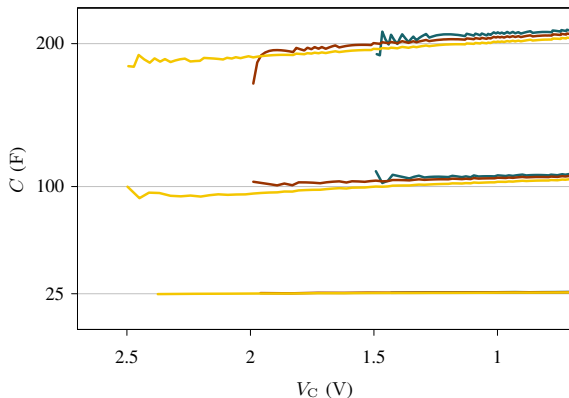
- ◆ Node powered by SuperCap
- ◆ $P_N = 51.2 \text{ mW}$ (radio on)
- ◆ $\eta \approx 85 - 91\%$ for $V_C = 1.5 - 2.5 \text{ V}$

Stand-alone Discharging: Voltage



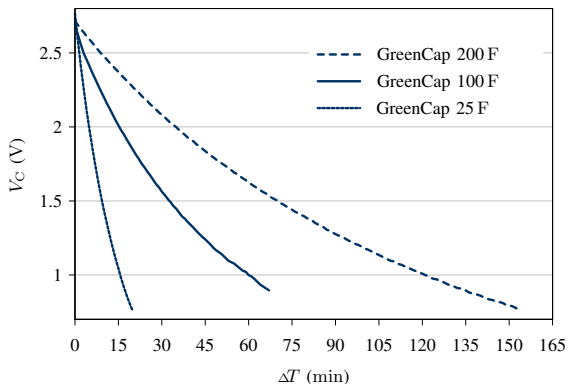
- Linear influence of C
- Exponentially decreasing V_C

Stand-alone Discharging: Capacitance



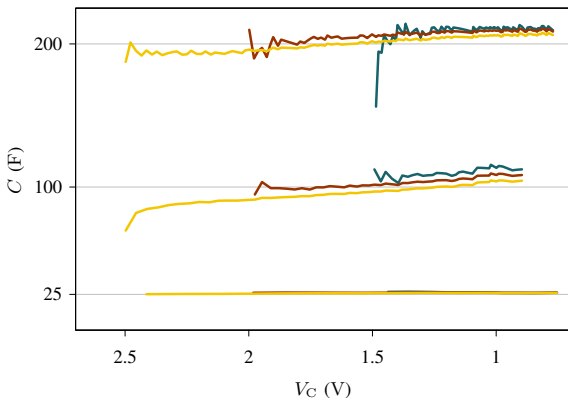
- C slightly increasing for smaller V_C
- Notable influence of sampling errors for $\Delta V_C < 0.1$ V

Combined Discharging: Voltage



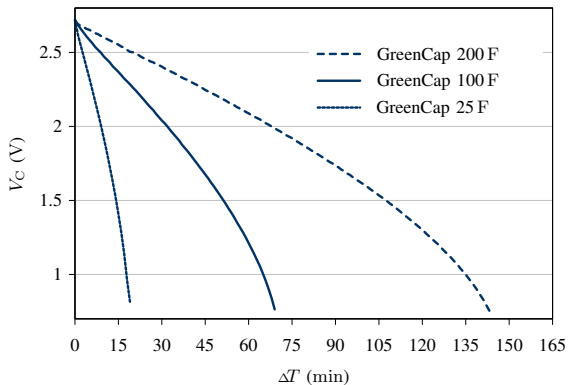
- Node power does not influence measurement significantly

Combined Discharging: Capacitance



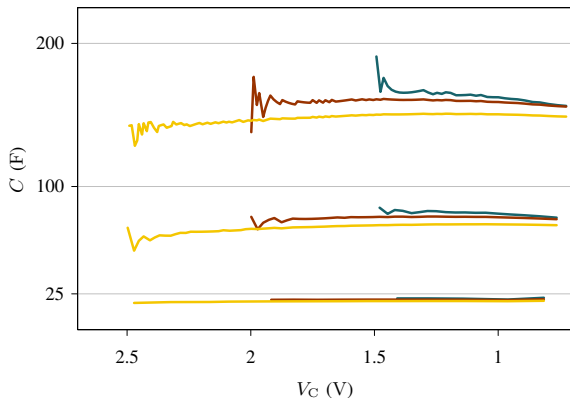
- Comparable results as before \Rightarrow approach feasible

Converter Discharging: Voltage



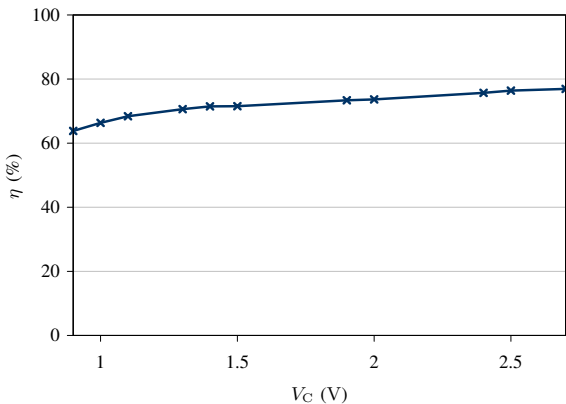
- Linear influence of C
- Square-root shape

Converter Discharging: Capacitance



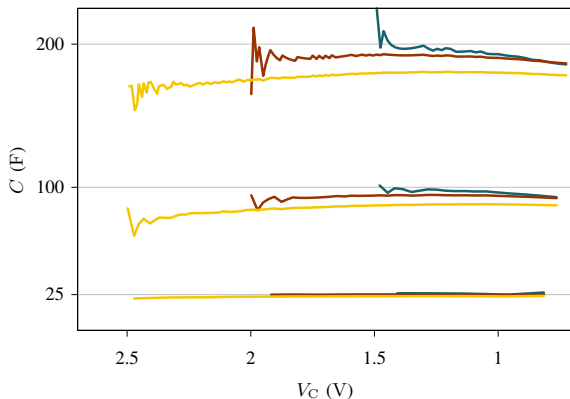
- Massive underestimation of C
- Heavy influence of V_{C0}

Circuit Efficiency



- Circuit efficiency considerably lower than converter efficiency

Converter Discharging: Capacitance Revisited



- Still heavy influence of V_{C0} : Large influence of leakage (discharging current lower than for resistance approach)

Summary

- Capacity estimation using resistance discharging is practicable
 - ◆ Low voltage drop needed ($\Delta V_C \approx 0.1 \text{ V}$)
 - ◆ Quick measurement possible ($\Delta T < 10 \text{ min}$ for $C = 200 \text{ F}$)
 - ◆ Good results
- Converter discharging has problems
 - ◆ Low current leads to high influence of leakage current
 - ◆ Good results only for low V_C
 - ◆ **but** can be used for efficiency determination

Roadmap for Online Capacity Determination

Initial estimation

1. Capacity determination
 - 1.1 Wait for appropriate V_C (e.g., 2 V)
 - 1.2 Put node into sleeping mode, enable resistance discharging
 - 1.3 Wait until discovering needed ΔV_C (e.g., 0.1 V)
 - 1.4 Estimate capacity C
2. Efficiency determination
 - 2.1 Disable resistance discharging
 - 2.2 Switch to high power mode
 - 2.3 Wait until discovering needed ΔV_C (e.g., 0.1 V)
 - 2.4 Estimate efficiency η

Maintenance Repeat estimation in service intervals or when performance of (lifetime) prediction degrades



Conclusion

Conclusion

- Refinement of lifetime prediction by
 - ◆ estimating and updating capacity
 - ◆ tracking actual circuit efficiency
- Proposition of two possible methods
- Analytical and practical evaluation on real hardware
- Starting point for self-calibration of an energy-harvesting power-supply



Future Work

- Convert knowledge into software
- Run field tests
- Refine algorithm
- Flip the coin: assess prediction of harvested energy





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Combined Discharging

▷ Back

Energy obtained from SuperCap

$$\Delta E_C = \frac{1}{2} C \left(V_{C0}^2 - V_C^2 \right) \quad (1)$$

Energy consumption of the converter

$$E_{\text{out}} = \frac{\Delta T \cdot P_N}{\eta} \quad (2)$$

Energy consumption of the resistor

$$E_R = \frac{V_{C0}^2}{R} \int_0^{\Delta T} e^{-\frac{2t}{RC}} dt = \frac{C V_{C0}^2}{2} \cdot \left(1 - e^{-\frac{2\Delta T}{RC}} \right) \quad (3)$$

With $\Delta E_C = E_{\text{out}} + E_R$

$$C \cdot \left(V_{C0}^2 \cdot e^{-\frac{2\Delta T}{RC}} - V_C^2 \right) = \frac{2 \cdot \Delta T \cdot P_N}{\eta} \quad (4)$$